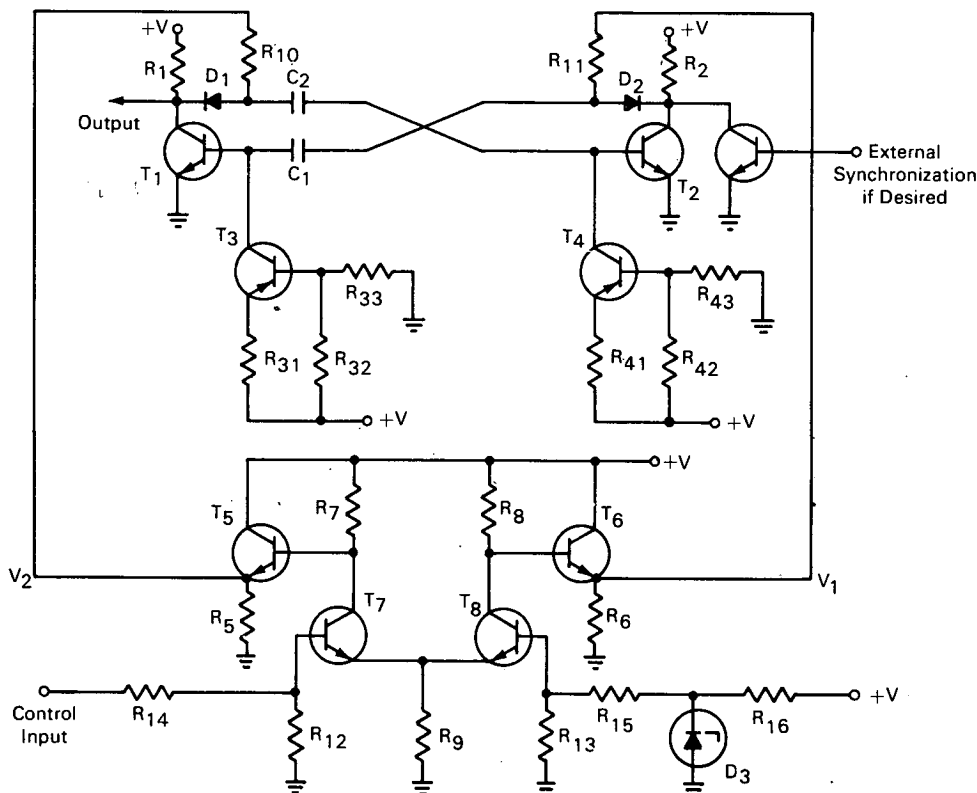


# NASA TECH BRIEF



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## Constant-Frequency, Variable-Duty-Cycle Multivibrator



### The problem:

To provide a pulse source of constant frequency with a duty cycle that is adjustable by an external input signal. Such a circuit would be most useful as a switching mode voltage regulator and could also find use as a switching source for a variety of control systems.

### The solution:

A circuit in which the above requirements are met and which may easily be synchronized by an external signal without interfering with the operation of the duty cycle control.

### How it's done:

Assume transistor  $T_2$  is on. The capacitor  $C_1$  has previously charged to potential  $V_1$ . With  $T_2$  on, capacitor  $C_1$  supplies a voltage  $-V_1$  to the base of  $T_1$ , thus holding it off. The current source formed by transistor  $T_3$  and its bias resistors  $R_{31}$ ,  $R_{32}$ ,  $R_{33}$ , causes this potential to increase linearly with time according to:

$$V_{beT_1} = -V_1 + \frac{I t}{C_1}$$

where  $V_{beT_1}$  is the base-emitter voltage of  $T_1$ ,  $I$  the current produced by  $T_3$ , and  $t$  time. When  $V_{beT_1}$  equals the "on potential" (i.e. forward bias potential)

(continued overleaf)

of  $T_1$ ,  $T_1$  turns on and  $T_2$  turns off due to cross coupling. Thus, the duration of the off time of  $T_1$  is given by  $t_1 = (V_{be} + V_1) \frac{C_1}{I}$  where  $V_{be}$  is the forward bias voltage drop of  $T_1$ . Similarly, the duration of the off time of  $T_2$  is given by  $t_2 = (V_{be} + V_2) \frac{C_2}{I}$ .

The total duration of a cycle of the oscillator is then  $T = t_1 + t_2 = \frac{C_2}{I}(V_1 + V_{be}) + \frac{C_2}{I}(V_2 + V_{be})$ . If  $C_1 = C_2$ , we have  $T = \frac{C}{I}(V_1 + V_2 + 2V_{be})$ .

Since  $V_1$  and  $V_2$  are derived from the differential amplifier  $T_5$ ,  $T_6$ ,  $T_7$ ,  $T_8$ ,  $R_5$ ,  $R_6$ ,  $R_7$ ,  $R_8$ ,  $R_9$ , the sum  $V_1 + V_2$  is constant. Thus, the frequency of operation is constant, but the duty cycle, or equivalently, the off time of  $T_1$ , is a linear function of  $V_1$  (see previous equation for  $t_1$ ).  $V_1$  is in turn a linear function of the control input signal. The duty cycle is thus controllable at constant frequency. The diodes  $D_1$  and  $D_2$  serve to decouple the charging of capacitors  $C_1$  and  $C_2$  from the power supply  $+V$ . Resistor  $R_{16}$  and Zener diode  $D_3$  provide a reference input so that the duty cycle is a function of the difference between the control input and the reference voltage of  $D_3$ .

The circuit may, therefore, also be used as a pulse duration modulator.

#### Notes:

1. The operation of the circuit is unchanged if a more complex differential amplifier is used instead of  $T_5$ ,  $T_6$ ,  $T_7$ , and  $T_8$ . Resistor  $R_9$  may be replaced by a transistor current source network similar to transistor  $T_3$ ,  $R_{31}$ ,  $R_{32}$ , and  $R_{33}$ . Any other constant voltage source may be used in place of  $D_3$  or it may be replaced by a resistor voltage divider network.
2. Inquiries concerning this innovation may be directed to:

Technology Utilization Officer  
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Greenbelt, Maryland 20771  
Reference: B69-10512

#### Patent status:

No patent action is contemplated by NASA.

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